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SECTION I

Features & Specifications

Magnavox television chassis CT 270, 271, 272, 273, 274 and 275 are 20 tube chassis. The CT 270, 272 and 274 include audio amplifiers and a speaker. The CT 271, 275 and 275 uses no audio amplifier or speaker, but is used in combination with a radio chassis audio system. However, as the audio output stage also acts as a voltage regulator, it appears in all models. All include direct viewing picture tubes. Special features are:

- * Capacity coupled video amplifier.
- * Intercarrier I F amplifiers for ease of alignment, simplified tuning, increased stability and freedom from the effect of oscillator drift.
- * Automatic Gain Control.
- * Magnalok type horizontal frequency control.
- * Blocking oscillator type vertical scanning.
- * Ratio detector sound detector.
- * Cathode modulation of picture tube.
- * Common chassis ground and B-

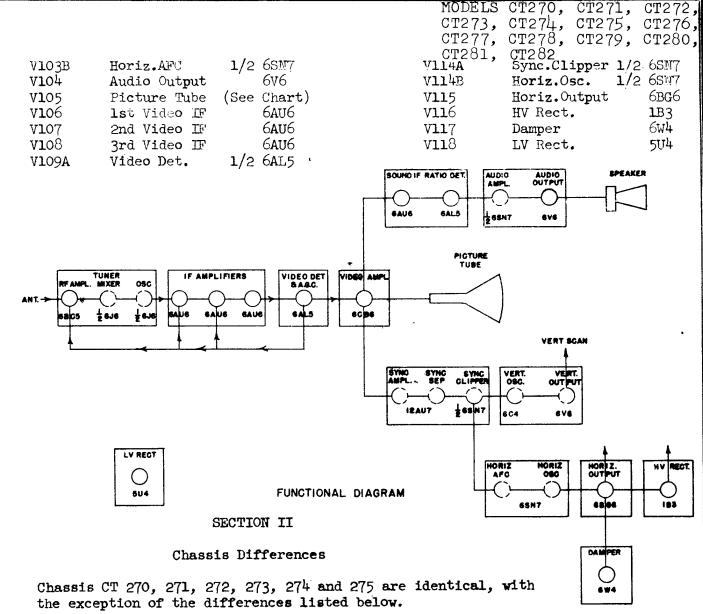
General Description

300 Ohm input------Impedance------Speaker coil 3.2 Ohms

185 Matts at 117V, 60CPS-----Power-----Audio 2.5 watts unlist.

Tube Complement

V1 V2A V2B V101	RF Ampl. Mixer Oscillator Sound IF Ampl	6BC5 1 / 2 6J6 1 / 2 6J6 6AU5 6AL5	V109B V110 V111A V111B V112	AGC Det. 1/2 6AL5 Video Ampl. 6CB6 Sync.Ampl.1/2 12AU7 Sync.Sep. 1/2 12AU7 Vert.Osc. 6C4
∀102	Ratio Det.	6AL5	V112	Vert.Osc. 6C4
V103A	1st Audio Ampi	1 1/2 6SN7	V113	Vert.Output 6V6



	Picture Tube	Туре	Audio Amplifiers
270	14BP4	Rectangular	Self Contained
271	14BP4	tt	None needed, Used with radio chassis
272	(16TP4 (16KP4	11	Self contained
273	(16RP4 (16XP4	11	None Needed, Used with radio chassis
274	16GP4	Round	Self Contained
275	11	***	None needed. Used with radio chassis

SECTION III

Operation and Installation

- 1. Plug the receiver into power outlet.
- 2. The AC switch is on the contrast control. Turn the receiver on by turning the CONTRAST OFF-ON control to the right about 1/2 turn.
- 3. Adjust the BRIGHTWESS control for suitable picture brilliance. This control adjusts the bias on the picture tube. If no light appears, it may be limited by incorrect adjustment of the ion trap magnet, which may be aligned as follows:

Advance the brightness control and adjust the ion trap until light appears on the screen. Then reduce the brilliance to a point near extinction by turning the brightness control counter-clockwise. Then readjust the ion trap until maximum brilliance is obtained. It may be necessary to turn the brightness control still further counter-clockwise.

- 4. Adjust the STATIOF SELECTOR to the desired channel number.
- 5. Adjust the FINE TUNING control for best picture quality. Do not tune for best sound as a distorted picture may result.
- 6. Adjust the CONTRAST control for the proper degrees of black, grey and white. This control adjusts the cathode degeneration on the video amplifier tube. With the brilliance at a point near extinction, adjust the control until sufficient picture detail appears without over-shading the light areas of the picture.
- 7. Adjust the VERTICAL HOLD control so the picture holds in vertical sync. If the picture "rolls" up or down, turn the vertical knob to the right so the roll is downward, then advance it to the left until the picture stops moving. Switching from channel to channel will not cause the picture to lose sync if the control is properly adjusted.
- 8. If the picture pulls out of sync in a horizontal direction, adjustment of the HORIZONTAL PHASING control (rear coil in the Magnalok transformer, rear panel or the instrument) and/or the HORIZONTAL SPEED coil (front adjustment of the transformer) may be necessary.
- 9. Adjust the FOCUS control so the lines making up the raster are clear and distinct.
- 10. Adjust the HEIGHT and VERTICAL LINEARITY controls so the picture just fills the mask and that linearity is uniform from top to bottom on all available channels. Adjustment of either the vertical hold, vertical linearity or height controls will slightly affect the other controls, and they may require a slight readjustment.
- 11. Adjust the HORIZONTAL DRIVE trimmer for the most drive consistant with good linearity. Turn it to a point where compression at the middle of the picture occurs, and then reduce it slightly.

- 12. Adjust the HORIZONTAL SIZE coil (top of chassis, adjacent to HV compartment) for proper width consistent with good linearity.
- 13. HORIZONTAL and VERTICAL CENTERING of the picture is accomplished in the conventional manner, using the controls on the rear panel. It it is impossible, loosen the small hex head screw on each side of the FOCUS MARNET and slide the magnet around for proper centering with no shadows. If this adjustment is unsatisfactory, loosen or tighten one or more of the three #1/4-20 adjusting screws on the focus magnet assembly. Readjustment of the ion trap may be necessary.
- 14. Adjust the HORTZONMAL LINEARUMY control for the most linear picture. CAUTION: There may be two positions of the slug which seems to provide proper horizontal linearity. Use the one which places the slug nearest all the way in.
- 15. Proper horizontal frequency adjustments can be made only after the other horizontal adjustments have been made. Note addenda, page 36.

Connect a weak signal of 100-1000 microvolts to the antenna terminals. Turn the horizontal hold control to one end of its range. Interrupt the signal by switching off channel. Turn the horizontal hold control to the other end of its range and again interrupt the signal. If the picture pulls out of sync. or shows any tendency to pull out, proceed as follows:

- a. Set the HORIZONTAL LOCK trimmer 3 turns out from tight.
- b. Connect a clip lead across the horizontal phasing coil and capacitor (terminals C and D of the Magnolok transformer) and adjust the horizontal speed coil (front of the transformer) so the picture holds over the entire range of the horizontal hold control when the signal is interrupted.
- c. Remove the clip lead and adjust the horizontal phasing coil so the picture holds over the entire range of the horizontal hold control when the signal is interrupted. CAUTION: If the horizontal phasing coil requires more than a slight adjustment, repeat the horizontal speed coil adjustment.

CABINET ANTENNA CONNECTIONS

A power line antenna and a cabinet antenna are included in this model. They may be used separately or together, depending on local conditions. Each should be tried on all available channels, and that antenna or combination giving the best results should be selected.

1. Connect the power line antenna lead (grey wire extending from the rear of the chassis) to first one and then the other terminal of the antenna terminal board, tune for all local television stations and note which gives the better results.

- 2. Remove the power line antenna lead from the terminal board, connect the cabinet antenna leads first one way and then reversed, tune for all available stations, and note the results.
- 3. Leave the cabinet antenna leads connected, then try the power line antenna to first one side of the terminal board and then the other. Reverse the cabinet antenna leads and repeat the two power line connections. Tune for all available stations in each condition (4) and note the results.
- 4. Select the one connection as described in 1, 2 or 3 that gives the best overall performance for all available stations.
- 5. Disconnect power line antenna and cabinet antenna when an outside antenna is ued.

PICTURE TUBE REPLACEMENT

- 1. From the rear of the cabinet, remove the back cover, the tube socket, the ion trap, and the high voltage anode connector.
- 2. The front glass assembly is held in position with a single horizontal support rail. In some models the rail is fastened to the glass support brackets with two small hex head screws, and in others the rail is secured to the cabinet sides with four small wood screws. Remove these screws and take off the rail. Exercise caution when the rail is removed that the glass does not drop of its own weight.
- 3. Loosen the nuts that hold the tube strap over the perimeter of the tube, and lift out the tube.
- 4. Install the new tube, replace the tube support strap, anode connector, ion trap, tube socket and the back cover. Make sure the strap is tight, as a loose tube will result in a distorted picture, and is subject to damage.
- 5. Be sure the deflection yoke is as far forward as it will go, that the grounding spring on the yoke makes contact with the grounding coating on the glass tube, and that the neck of the tube is parallel to the deck of the chassis. The shoulder of the tube should be tight against the kinescope retaining ring support.
- 6. Replace the safety glass and support rail.

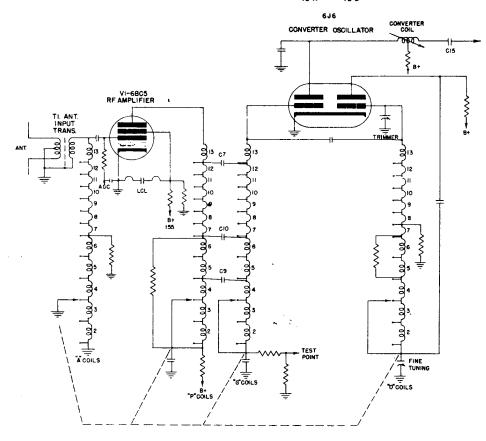
CIRCUIT DESCRIPTION

SECTION IV

The Magnavox M-1 tuner consists basically of two tubes, four sets of coils and a four section, twelve position wafer switch. The tubes are V1, the RF amplifier 6BC5, and a dual triode 6J6, V2-A and V2-B. One triode section, V2-A is the converter or mixer, and V2-B is the oscillator.

The antenna input transformer Tl is provided to match a balanced 300 ohm antenna to the tuner. The ANTENNA COHS A-13 to A2 are in the grid circuit of the RF amplifier stage, and are grouped around the rear wafer of the unit. The RF amplifier PLATE coils P-13 to P-2 are mounted on the second wafer from the rear. The converter GRID coils G-13 to G-2 are on the third wafer from the rear. The OSCHLATOR coils 0-13 to 02 are on the front wafer. In each case, coil 13 is tuned for channel 13; coils 13 plus 12 are for channel 12; coils 13 plus 12 and 11 are for channel 11, etc. The following simplified diagram outlines its design and operation.

MODELS CT270, CT271, CT272, CT273, CT274, CT275, CT276, CT277, CT278, CT279, CT280, CT281, CT282



SIMPLIFIED MAGNAVOX M-I TUNER DIAGRAM

Antenna coils A-13 to A-2 are tuned to their respective channels, with a band pass pattern sufficiently wide to accept all the information therein, and with sides of the pattern sufficiently steep so as to reduce signals on adjacent channels. The same applies to coils P-13 to P-2. The L-C-L unit is to provide regeneration on the higher channels to overcome degeneration due to tube loading. Under weak signal conditions, the RF amplifier stage is capable of a gain of 15 times its input. In the presence of a strong signal, the automatic gain control bias on the grid limits its output to a fraction of its input.

The oscillator is tuned on a frequency 25.75 mc. higher than the video carrier, and when hetrodyned with the carriers in the converter, produces a modulated video IF signal of 25.75 mc. and a modulated sound IF signal of 21.25 mc. These are applied to the IF amplifiers through C-15. L-4 in the mixer plate circuit is series tuned, with the tube capacities of the converter and first IF tubes, to prevent oscillator voltage from being applied to the IF amplifiers.

The fine tuning control, a variable dielectric capacitor, shifts the oscillator frequency about 1.5 mc. on channel 2, and relatively more on the other channels up to 13, where it is about 4.5 mc. The trimmer capacitor, accessible from the front of the chassis by removal of the channel scale and fine tuning knobs, is provided to compensate for dif-

ferent tube capacities when the oscillator tube is replaced. It will shift the oscillator frequencies on all channels in the same direction (depending on the direction of adjustment) in the event that tuning occurs too close of the end of the fine tuner rotation one one or more channels.

The oscillator signal is applied to the converter grid through a capacitor from the oscillator plate. The amplified RF signal is applied to the converter grid coils through capacitors C-7, C-9 and C-10, depending on the position of the channel selector switch. These capacitors are used to provide adequate bandpass coupling. The converter stage is capable of a gain of about 5 times its input.

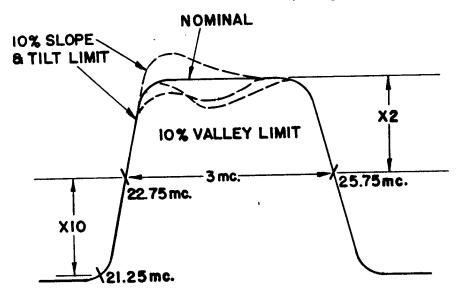
A test point is provided that is accessible through a hole in the right side of the tuner cover for connecting an oscilloscope to be used in alignment of the antenna, RF and converter coils. A complete schematic diagram of the 700340 RF unit, alignment procedure and service instructions are included in a later chapter.

The tuner is capable of a gain of about 75 times from the antenna to the first IF amplifier grid.

The IF amplifier strip passes both video and sound IF signals to the common detector. It consists of three 6AU6 tubes, V-106, V-107 and V-108, coupled by IF transformers T-102, T-103 and T-104. The video IF signal is 3 mc. wide, extending from the picture IF carrier of 25.75 mc. to 22.75 mc., which is 6db down.

The primary and secondary winding of the IF transformers are bi-filar wound (in the same direction), and coupling between stages is by the capacity and inductance between windings. Each transformer is tuned with an adjustable iron core.

The converter coil and the three IF transformers are appropriately loaded, stagger tuned, and resonate at various IF frequencies to provide maximum gain for the desired bandpass characteristic, thus eliminating the need for over-coupled transformers, traps and filters.



IDEAL OVERALL IF CURVE

The converter coil and T-104 are high Q coils, and contribute to the steep sides of the overall bandpass curve. T-102 and T-103 are also high Q coils, but are sufficiently loaded to pass a broad band of frequencies, and contribute to the overall IF gain.

Each IF stage has an LC isolation network in its filament circuit to prevent interaction between stages and coupling of the IF voltages into the stages of the receiver through the heater-capacitance of the tubes in the common filament line.

AGC bias is applied to V-106 and V-107, and control their gain inversely proportional to the applied signal strength. V-108 is operated at fixed bias. The output of the IF system has a rising characteristic as the input level increases.

Complete alignment instructions are included in a later chapter.

The video detector, V-109A consists of 1/2 of a 6AL5. On negative cycles of the IF carriers, the diode conducts and current will flow through R-125. This current develops a voltage across R-125, which varies with modulation of the IF carrier. The high frequency IF component is bypassed to ground through C-122.

The same detector operates as a mixer for the video IF and sound IF signals. The resultant FM and AM 4.5 mc. beat appears across R-125 along with the video signal. As the IF carriers result from crystal controlled frequencies at the television transmitter, their resultant beat is much more stable than if the sound IF were produced by beating the sound carrier with a local oscillator.

The video amplifier has an essentially flat characteristic up to 3 mc., and then tapers off to 0 at 5 mc. L-108, L-109, L-106and L-111 provide peaking necessary for this broad-band-output, which is adequate for both the video signal and the sound IF. It has a gain of 30 times, and with a normal input of 2 volts, will present 60 volts video modulation to the kinescope.

The contrast control is in the cathode return circuit, and serves as a direct control over the input to the tube. As the output from the detector is in a negative direction, noise pulses drive the video amplifier beyond cutoff, so it also acts as a noise limiter.

The picture tube operates on a second anode potential of 12.5 KV. As the output from the video amplifier is in a positive direction, the signal is applied to the kinescope cathode through C-129. This is equivalent to grid modulation with a negative signal. R-178 and R-177 maintain a DC voltage on the cathode. The brightness control adjusts the DC bias on the grid, and governs the emission from the electron gun, which produces the brilliance.

The series parallel resonant sound IF pick-off coil, composed of L-107 and C-126 samples the 4.5 mc. signal from the video amplifier plate, and it is applied to V-101, the sound IF amplifier, where it is amplified sufficiently to drive the ratio detector. The latter stage accepts the frequency modulated audio signal and provides a high degree of rejection to amplitude modulation.

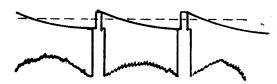
MODELS CT270, -1, -2, -3, -4, -5, -6, -7, -8, -9, 80, 81, 82

The audio amplifier and audio output stages are conventional, except for the DC potentials present on the output tube, which are 360 on the plate, 155V on the cathode, 230V on the filament and only a slightly negative voltage on the grid. A separate filament winding in the power transformer is used for V-104 and V-117.

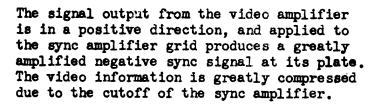
The other section of the 6AL5, V-109B serves as the Automatic Gain Control Rectifier.

Its cathode is maintained at approximately 1.5 volts positive by direct connection to the cathode resistor of V-108. The modulated RF envelope is applied to its plate. Because its plate resists being driven positive in respect the cathode, the resultant current causes C-158 to be charged to a potential that causes the AC axis of the RF envelope to fall below zero to an extent that the positive sync pulses of the envelope exist above zero by a voltage slightly greater than the bias on the ACC diode cathode. C-120 then becomes charged through R-122 as the resultant of the ACC diode plate existing at an average level below zero. R-122 and C-120 provide a sufficiently long line constant to eliminate ripples in the ACC Bus.

The time constant of C-158 and R-151 is such that the positive peaks of the RF envelope are able to follow the horizontal sync pulses.

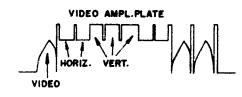


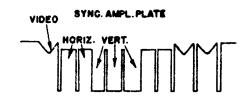
Thus far, no mention has been made of the vertical and horizontal sync pulses, which appear on the video carrier, through the IF amplifiers, detector and video amplifier, between scanning lines of video modulation. Their appearance at the picture tube grid is always during retrace time, and cuts the beam to blanking level at those intervals.

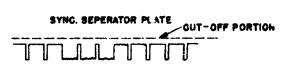


The sync separator stage follows, which is biased high enough to pass only the sync pulses. Everything else is below the cut-off portion of the tube characteristic. Here again, the output is in a negative direction.

Noise and distortion if any, rides the peaks of the sync pulses. These pulses are applied to the sync clipper stage, which has sufficient cut-off to remove the peaks and present clean, square wave pulses to the deflection stages.









CT274, CT275, CT276, CT277, CT278
CT279, CT280, CT281, CT282

irectly to the AFC and horizontal

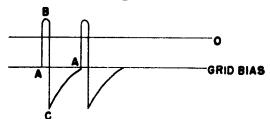
MODELS CT270, CT271, CT272, CT273

The sync clipper output is fed directly to the AFC and horizontal stages through C-135, which is sufficiently low in value so as to pass only the high frequency horizontal sync pulses. The same signal is applied to the vertical oscillator through RC-101, a filter network which passes only the low frequency vertical pulses.

The vertical oscillator and output stages provide a sawtooth current wave of proper shape and frequency through the deflection yoke for vertical scanning of the picture tube.

The first section, the 6C4 (V-112), is a blocking oscillator and diacharge circuit. During non-conduction, the grid voltage is negative with respect to the cathode. The plate draws current during discharge at a rate determined by the setting of the height control (R-148). Due to the coupling of the oscillator transformer, there is a corresponding voltage rise on the grid (A to B on following curve). When the grid becomes more positive than the cathode, it draws grid current. This quickly charges C-136, which drives the grid negative and cuts off the plate current

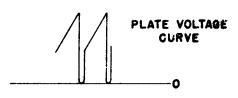
GRID VOLTAGE CURVE



(see B to C). Then the charge on C-136 is slowly discharged through the vertical hold control R-144, and R-150, which allows the grid voltage to slowly rise to its normal bias (see C to A). Then plate current begins to flow again, and the cycle is repeated, at a frequency depending on the rate of C-136 discharge, which is controlled by the setting of R-144.

The frequency is set at slightly slower than 60 c.p.s. During the charging period of C-136 (C to A), the vertical sync pulse is applied, just before it would "trip" in its free-running cycle. The magnitude of the sync pulse is sufficient to drive the tube into conduction, and therefore controls the frequency of oscillation.

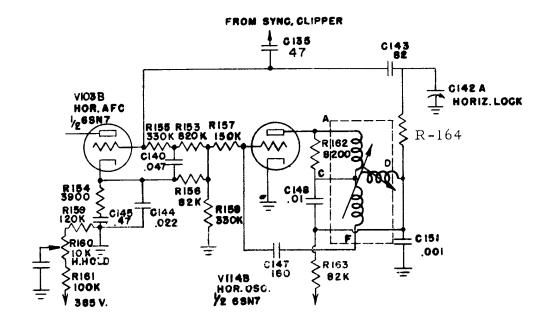
The oscillator plate voltage consists of a series of sawtooth waves. A sharp negative pulse appears because R-146 prevents C-137 from complete discharge to ground when the tube is conducting. When the tube becomes non-conducting, the plate voltage does not have to rise slowly from zero, but instead immediately rises to a value determined by the value



of R-146, principally because the instantaneous changes do not charge C-137, The plate voltage then slowly rises from this value as C-137 charges. Adjustment of the height control (R-148) varies the amplitude of the sawtooth voltage by controling the rate at which C-137 can charge.

The voltage present on the plate is of the basic shape required to produce a sawtooth of current in the deflection coil, and is amplified by the 6V6 output stage, V-113.

V-114B is a blocking oscillator. Its frequency is controlled by adjustment of the frequency coil (front adjustment of the Magna-lok transformer). The horizontal speed coil (rear adjustment) is included to stabilize oscillation, especially in the presence of noise. The bias on V-114B is obtained through a voltage divider network from the cathode of V-103B, and any bias variation will change the oscillator frequency.



The sawtooth wave is fed to the grid of the automatic frequency control stage V-103B, through R-164, along with the horizontal sync pulses from C-135. C-143 and the horizontal lock trimmer comprise a capacity voltage divider which controls the amplitude of the fed back sawtooth waves and sync pulses.

V-103B is biased to nearly cut-off. When the oscillator frequency is the same as that of the sync pulses, the pulse rides on the crest of the fed back wave, and is above the cut-off level of the tube. It is then at maximum width, V-103B conducts during its appearance at that level, and cathode capacitors C-144 and C-145 charge to a DC potential proportional to plate conduction time. This voltage is used to bias the oscillator grid. In this example, the bias is of such value so as to sustain oscillation at its present frequency.

However, if the oscillator frequency changes, due to component warm up, etc., the sync pulse will reach the AFC grid slightly ahead or behind the saw tooth wave. Part of the pulse will ride the slope of the wave and will appear below the tube cut-off level. The portion of the sync pulse on the crest of the saw tooth wave will be at less than maximum width, and will limit tube conduction proportionately. V-103B cathode condensers will charge to a different voltage than in the previous example, the bias on V-114B will be changed, and the oscillator frequency will be pulled into sync.

The horizontal hold control will vary the frequency about the mean horizontal sweep by varying the AFC tube plate voltage.

METHOD OF OBTAINING REQUIRED DC VOLTAGES

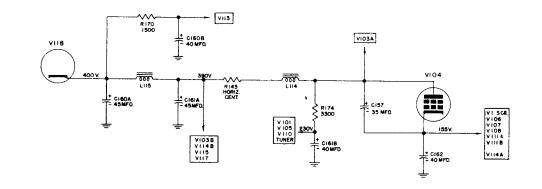
This chassis contains a highly efficient voltage divider network, which eliminates the necessity for power consuming and heat producing bleeder resistors.

The input condenser C 160A is at 400 V, which is applied to the vertical amplifier stage. The first filter choke L115 drops it to 390, which is used for the Horizontal AFC, Horizontal Oscillator, Horizontal Output and Damper stages. The second filter choke drops it to 385 V, which is needed for the audio amplifier and audio output stages. R 174 drops the voltage to the required 230 for the sound IF, picture tube grid, video amplifier and the tuner.

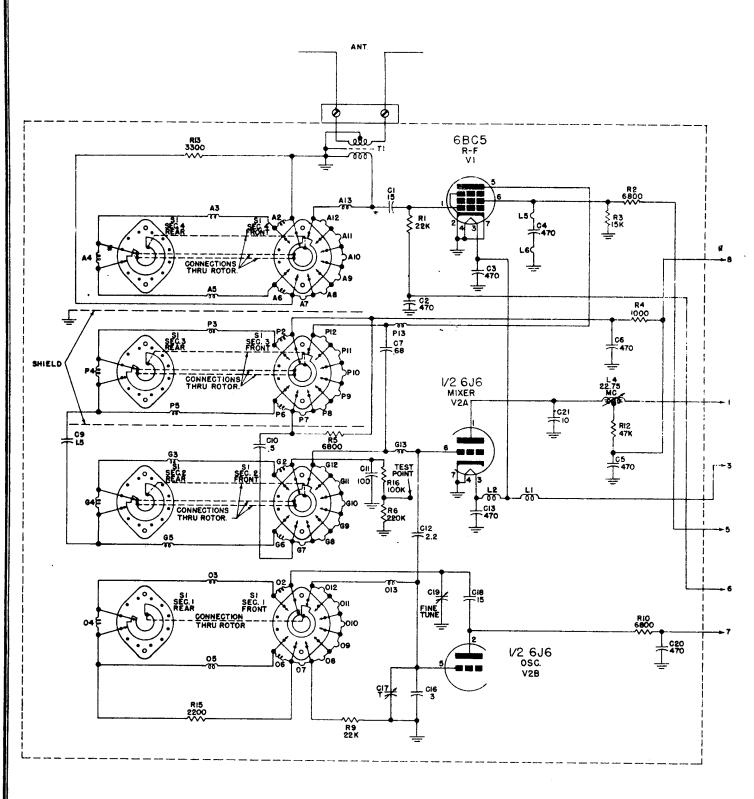
The audio output stage is also a voltage regulator. The RF amplifier screen, 1st, 2nd and 3rd IF amplifiers, sync amplifier, sync separator, sync clipper comprise its cathode load, and are maintained at 155V.

The damper tube produces boosted $B \neq in$ the conventional manner, which is applied to the horizontal output plate and the picture tube screen.

A separate filament winding on the power transformer for V 117 and V 104 is required to eliminate excessive heater cathode voltage.



MODELS CT270, CT271, CT272, CT273, CT274, CT275, CT276, CT277, CT278, CT279, CT280, CT281, CT282



SECTION 5

· R-F AND I-F ALIGNMENT

TEST EQUIPMENT. For proper television alignment, it is recommended that the following test equipment be available.

1. R-F Sweep Generator meeting the following requirements:

a. Frequency

18 to 30 MC 40 to 90 MC

170 to 225 MC

- o. Sweep width adjustable from 1 to 12 MC
- c. Cutput at least .1 volt maximum.
- 2. Cathode Ray Oscilloscope Wide band vertical deflection.
- 3. Electronic Voltmeter with high voltage probe.
- 4. 500,000 ohm potentiometer with clip leads and 3 volt battery.
- 5. Signal generator to cover the following frequency ranges.

I-F range from 19.75 MC to 27.25 MC.

R-F range from 55.25 MC to 87.75 MC and 171.25 MC to 215.75 MC

R-F UNIT ALIGNMENT

Oscillator alignment should be made only when the fine tuning control tunes in the extreme clockwise or counter-clockwise position, or if it will not tune at all within its tuning range.

1. TOUCH UP OSCILLATOR ALIGNMENT

a. If some channels do not tune at all, or not near enough to the center of the fine tuning range, adjust the oscillator trimmer for the best compromise tuning on all channels. If, for example, all channels tune to one side of the control, adjustment of the trimmer will bring them all toward the center. However, if some channels tune at one end of the control and others tune at the other end, adjustment of the trimmer will bring the former to the center of the tuning range and will force the latter beyond the range of the control. In this case, the oscillator coils must be aligned individually as follows:

Channel No.	Band Width (MC)	Picture Carrier	Sound Carrier	Local Oscillator
2 3 4 5 6 7 8	54-60 60-66 66-72 76-82 82-88 174-180 180-186	55.25 61.25 67.25 77.25 83.25 175.25 181.25	59.75 65.75 71.75 81.75 87.75 179.75	81 87 93 103 109 201 207
9 10 11 12 13	186-192 192-198 198-204 204.210 210.216	187.25 193.25 199.25 205.25 211.25	191.75 197.75 203.75 209.75 215.75	213 219 225 231 237

2. COMPLETE OSCILLATOR ALIGNMENT

- a. Connect a sweep generator to the antenna terminals.
- b. Connect an unmodulated signal generator to the antenna terminals and tune it to the channel 13 video carrier frequency. TV signals may be used.
- c. Loosely couple an unmodulated signal generator to the first IF grid, and tune it to 25.75 mc.
- d. Connect an oscilloscope across the video detector load resistor.
- e. Turn on the receiver and set the straight edge of the vernier to straight up position (mid range). After the receiver has warmed up and if there are two markers on the scope, determine which rotation of the vernier causes the markers to move closer together. If it is clockwise, move the fine tuner back to the straight up position and close the channel 13 oscillator coil. If it is counter-clockwise, open the coil. Adjust for zero beat. See note 1.
- f. Tune the signal generator connected to the antenna terminals to the video carrier frequency for channel 12, switch the tuner channel selector switch to its channel 12 position and adjust the channel 12 oscillator loop for a zero beat of the markers.
- g. Repeat the procedure for channels 11 through 2, and recheck 13 through 7. The high channel loops are adjusted by moving them up or down and the low channel coils by spreading or pushing together.

ANTENNA, RF AMPLIFIER AND CONVERTER ALIGNMENT

1. There are three variables which will give the desired pattern to be ap-

NOTE 1. Usually, if the oscillator is off in the same direction for the high channels, adjustment of Channel 13 will bring in all high band channels. The same applies to all low channels if they are off in the same direction and Channel 6 is re-adjusted.

plied to the IF amplifiers. If the ANTENNA COIL (wafer nearest the rear of the tuner) is first adjusted, then the RF COIL (second wafer from the rear), and finally the CONVERTER COIL (third wafer from the rear), the proper pattern will appear.

These coils must be very carefully adjusted with only slight movement, particularly the converter coil. The RF wafer has B/ on it and should not be touched with the hand or a metal tool. The coils around the first (oscillator) wafer are properly adjusted and should not be touched.

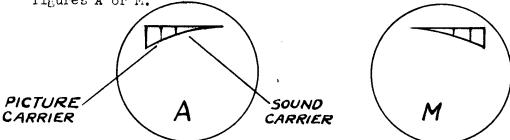
The following alignment instructions are for use where over-all alignment is necessary. When a partial touch-up is necessary, follow the test patterns through the outline until the one is found that most nearly matches the curve shown on the oscilloscope, and then complete the rest of the alignment procedure.

2. HOOKUP

- a. Attach a VTVM to the AGC terminal (green wire, bottom terminal on left side) of the tuner and chassis ground. Using battery and 500K potentiometer, adjust the bias for -2V.
- b. Connect a sweep generator to the antenna terminals, using a balanced resistor network if the generator has an unbalanced output cable.
- c. Loosely couple a signal generator to the antenna terminals, either by wrapping a few turns of insulated wire around the terminals and connecting the signal generator to it, or through a small capacitor.
- d. Connect an oscilloscope, through an isolating resistor of about 47 K ohms, to the test point, which is accessible through the hole in the right side of the tuner, top side of chassis.

3. ALIGNMENT, MODEL M TUNER

- a. Turn the channel selector switch on the tuner and the sweep generator to Channel 13.
- b. Adjust the signal generator to first the video carrier frequency and then to the sound carrier frequency for channel 13. If the coils are badly adjusted, the pattern on the scope will appear as in figures A or M.



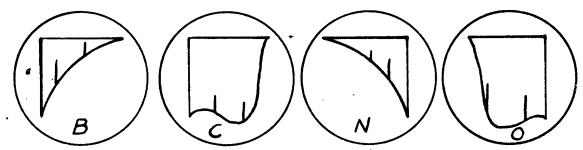
The meaning of A is that some coils must be opened and of M that some must be closed. When the coils are far off from their proper tuning, all three have their main effect on pattern height, so tune for increasing height of pattern. MODELS CT270, -1, -2, -3, -4,

-5, -6, -7, -8, -9, 80, 81, 8

c. As a coil is changed, one side of the picture will get higher and a hump will start to move across the screen. When starting with a pattern like A, first B, then C, will appear as a coil is opened.

When starting with a pattern like M, first N, then O will appear as a coil is closed.

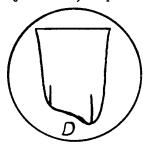
When the hump which is due to tuning a coil is between the markers, it indicates that this coil is very close to where it should be tuned. Leave it as it is and go immediate to another coil.

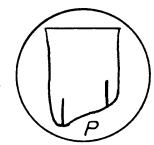


d. The lower side of the pattern shows what to do to the next coil. In C, the left side is lower, which shows that the next coil should be opened.

In O, the right side is lower, which shows that the next coil should be closed.

After this adjustment, a pattern like D or P should be on the screen.





The fact that there are now two humps in the pattern shows that both converter and RF wafers are nearly right. If the pattern height is near what it usually is with the scope setting that you have been using, the antenna wafer is probably near correct, but it should be adjusted a bit to see if pattern height improves.

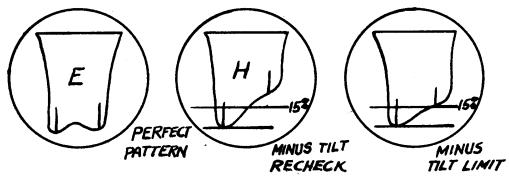
- 4. ROUGHLY ALIGN THE LOW BANDS.
 - a. Now that Ch. 13 is roughly aligned, go immediately to Ch. 6 and roughly align it for the pattern in L. Do the same for Channels 5, 4, 3, and 2.
- 5. FINAL ALIGNMENT OF CHANNELS 13 DOWN TO 2.
 - a. Now go back to Ch. 13. Do it the easy way on the tuner by going clockwise one step from 2.

MODELS CT270, CT271, CT272, CT273, CT274, CT275, CT276, CT277, CT278, CT279, CT280, CT281, CT282

- b. Align the Ch. 13 coils for the highest and narrowest picture you can get. You will notice that the narrower the pattern is, the higher it is also. The narrowest pattern you can get should be like E. Then align channels 12 down to 2, trying for a flat pattern with markers on the peaks in each case.
- c. You will be able to move the pattern to right and left by using the RF wafer. Tilts may be corrected by adjusting antenna wafer, converter wafer, or both of them. The adjustment should be used which gives best pattern height. Notice that although the antenna and converter wafers both control tilt, they have opposite effects on it. Opening an antenna low channel coil or pushing down an antenna high channel loop will bring down the right side of a pattern. Closing a converter low channel coil or pulling up a converter high channel loop will do the same thing.

Closing an antenna low channel coil or pulling up an antenna high channel loop will bring down the left side of a pattern. Opening a converter low channel coil or pushing down a converter high channel loop will have the same effect. Again, the adjustment to use is the one which will give the highest pattern.

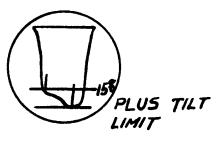
- 6. HOW TO CORRECT THE PATTERNS FOR RIGHT CENTERING AND RIGHT TILT.
 - a. Correcting for minus tilt.



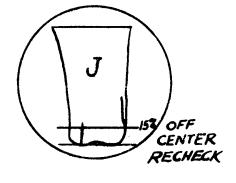
For minus tilt as in H, open a low channel coil, or push down a high channel loop of the antenna wafer. It can also be corrected by closing a low channel coil or pulling up a high channel loop on the converter wafer. It should be done in the way that gives best pattern height. You should always try to get a flat pattern if it can be done.

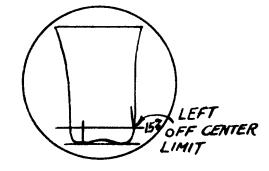
b. Correcting for plus tilt.





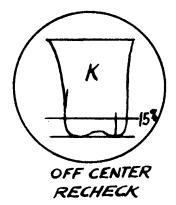
c. Correcting for off center left.

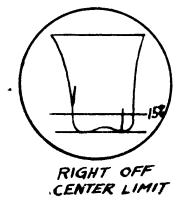




For a pattern like J, which is off center left, open a coil, or push down a loop on the RF wafer. Adjust until the pattern is symmetrical about the markers, that is until the markers are in the center of the pattern, instead of over to one side. Always try to get the markers as near centered as you can.

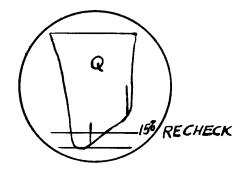
d. Correcting for off center right.

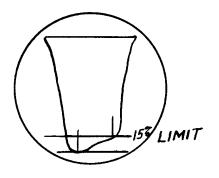




For a pattern like K, close a low channel coil or pull up a high channel loop on the RF wafer. Adjust until the pattern is symmetrical about the markers.

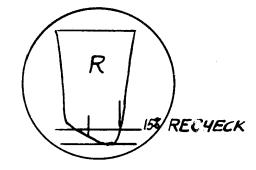
e. Correcting combination faults.
1. Off center left and minus tilt:

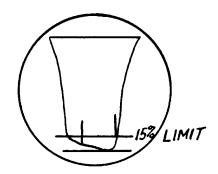




Note that Q is a combination of J and H. First correct the RF wafers for centering as in section C, then correct for tilt as in section A.

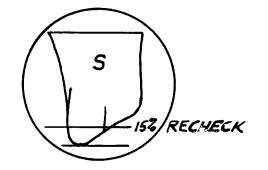
2. Off center left and plus tilt:

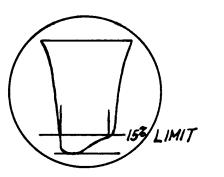




Note that R is a combination of J and I. First, correct the RF wafer for centering as in section C, then correct for tilt as in section B.

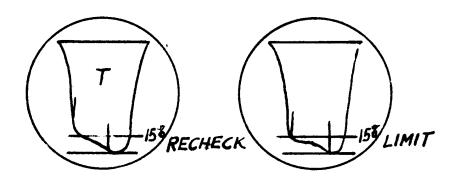
3. Off center right and minus tilt:





Note that S is a combination of K and H. First, correct for centering as in section D, then correct for tilt as in section A.

4. Off center right and plus tilt:



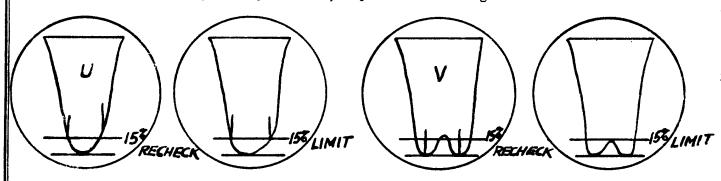
hote that T is a combination of K and I. First correct for centering as in section D, then correct for tilt as in section B.

Don't try to make every channel come in by adjusting one coil. Usually, at LEAST TWO MAFERS HUST BE ADJUSTED TO GET THE RIGHT PATTERN.

- 7. HOW TO CORRECT ROULD BOTTOM AND DEEP VALLEY.
 - a. For RB as in U, open a low channel coil or push down a high channel loop on both the antenna and converter wafers. Then correct for off center by adjusting the kF wafer.

If that doesn't correct it, try closing a low channel coil or pulling up a high channel loop on both antenna and converter wafers. Then correct for centering by adjusting the RF wafer.

b. For deep Valley as in V, try the same thing as:



for RB, that is, moving antenna and converter in one direction and RF in the opposite direction. Again, if moving the three coils in first trail doesn't correct it, then try the second way, just as for RB.

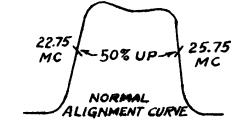
MODELS CT270, CT271, CT272, CT273, CT274, CT275, CT276, CT277, CT278, CT279, CT280, CT281, CT282

Video I.F Alignment

- a. Adjust the bias on the first & second IF grids at -3 volts by connecting a 3 volt battery from the grid return of either tube to ground,
- b. Connect the IF Signal Generator to the converter grid wafer (through the hole provided in the bottom of the tuner shield) and chassis ground.
- c. Connect an oscilloscope to the high side of the detector load resistor R-125, 4700 ohms (junction of R-125 and L-106), and to chassis ground.
- d. Tune converter coil T-102 for maximum at 22.9 mc. The purpose of the oscilloscope is to observe the demodulated signal and thus make sure that no overload condition develops because of too much signal input. This condition, of course, is evidenced by no increase in output voltage for an increase in signal, and a badly distorted wave form.
- 6. Tune first I F coil T-102 for maximum at 23.4 mc.
- f. Tune second I-F coil T-103 for maximum at 25.2 mc.
- g. Tune third I-F coil 'I-104 for maximum at 25.3 mc.
- h. Remove the I-F signal generator, and connect the R-F sweep generator to the antenna terminals.
- i. Tune the receiver and R-F sweep generator to channel 6.
- j. The I-F frequencies can now be retouched if necessary to obtain a standard response curve, with the picture carrier at 50% of maximum response.

The IF marker may be injected by connecting the IF signal generator to the tuner frame and turning the gain high enough so it will appear on the curve without causing overload

The converter coil T-102 will correct the low frequency side, as will T-103 and T-104 correct the high frequency side when necessary. Proper positioning of the markers on the response curve is important since no traps are used.



	Ref. No.	Part No.	Description
k. All twelve channels should be checked for alignment. The nominal	M1 03	2601.80 1	Ratio Detector Transformer
curve shall be flat with respect to the base line of the sweep. The	T101	360482-1 360461-1	lst Video IF Coil
tilt limit shall be 10% with respect to the flat portion of the curve.	T102		
	T103	360461-1	2nd Video IF Coil
	T104	360461-1	3rd Video IF Coil
SOUND IF ALIGNMENT	T10 5	320030-6	Vertical Blocking Oscillator Transformer
	T106	320056-2	Vertical Output Transformer
Test Equipment Required:	T10 7	320055-1	H.V. Transformer
. oo aquapmon noque ou	T108	300059-1	Power Transformer
4.5 mc. signal generator.	T109	360479-1	Horiz. Sync. & Osc. Reactor
	-20)	500.17	
Voltohmyst type VTVM.	RC101	250186-1	Printed Circuit
	HOTOL	180475-1	Fuse 1/4 Amp.
a. Connect the signal generator to the video amplifier grid (pin 1 of		1004/5=1	ruse 1/4 Amp.
V 110) and chassis.		a(a).0c 3	Thomas Marina
		360485-1	Focus Magnet
b. Connect probe of VTVM to the high side of detector load resistor R-105,		•	
6800 ohms (at pin 7 of V-102), and negative lead to chassis.	C101	250187-50	Capacitor, Mica, 56 mmf., 500 V.
	C102	250175-1	Capacitor, Ceramic, 5000 mmf., 450 V.
c. Turn contrast control fully counter-clockwise.	C104	250159-90	Capacitor, Mica, 470 mmf., 500 V.
At where Assess Min a Assess an Friends Assessed Appendix A.	C105	250088-46	Capacitor, Ceramic, 1000 mmf., 350 V.
d. Adjust signal generator to 4.5 mc.	C106	250175-2	Capacitor, Ceramic, 10,000 mmf., 450 V.
d. Adjust signal generator to 4.5 mc.	C107	270027-4	Capacitor, Electrolytic
	· ·	250 1 75 - 2	Capacitor, Ceramic, 10,000 mmf., 450 V.
e. Tune slug in 4.5 mc. sound take off transformer (L-107) for maximum	C103		
reading.	C109	250175-2	Capacitor, Ceramic, 10,000 mmf., 450 V.
	C110	250205-11	Capacitor, Paper, .047 mfd., 200 V.
f. Tune ratio detector primary (bottom slug of T-101) for maximum reading.	C111	250 1 75 - 3	Capacitor, Ceramic, 1500 mmf., 450 V.
	C112	250203-6	Capacitor, Paper, .0068 mfd., 600 V.
CAUTION: Keep the 4.5 mc. signal input low enough so the VTVM reading	C113	250 1 75 - 3	Capacitor, Ceramic, 1500 mmf., 450 V.
does not exceed -5 V, to prevent overload and misalignment.	C114	250175-3	Capacitor, Ceramic, 1500 mmf., 450 V.
does not exceed a) it so brevelle everyone und urbarreness	C115	250175-3	Capacitor, Ceramic, 1500 mmf., 450 V.
Demons MMMM compactions and compact magazine load to the impetion of D 105	C116	250175-3	Capacitor, Ceramic, 1500 mmf., 450 V.
g. Remove VTVM connections and connect negative lead to the junction of R-105	C117	250175-3	Capacitor, Ceramic, 1500 mmf., 450 V.
and R-129 (both 6800 ohm 5%), and the probe to the junction of R-103 (220	•	250175-1	Capacitor, Ceramic, 5000 mmf., 450 V.
ohms) and C-104 (470 mmf). Adjust secondary of ratio detector (top slug	C118		Capacitor, Ceramic, 1500 mmf., 450 V.
of T-101) for zero reading.	C119	250175-3	
	C150	250205-13	Capacitor, Paper, .1 mfd., 200 V.
If the secondary is too far out of alignment, (4 or 5 turns either di-	C121	250159 - 82	Capacitor, Mica, 100 mmf., 500 V.
rection) repeat peaking of primary as described in steps b, e and f."	C122	250207 - 3	Capacitor, Ceramic, 10 mmf., 450 W.
	C125	250187-19	Capacitor, Mica, 75 mmf. 5%, 500 V.
•	C126	250164-10	Capacitor, Ceramic, 1.5 mmf., 500 V.
SOUND IF ALIGNMENT, USING AIR SIGNAL	C128	270027-14	Capacitor, Electrolytic
poorte it terrorisative operior time pronum	C129	250203-11	Capacitor, Paper, .047 mfd., 600 V. Capacitor, Ceramic, 1500 mmf., 450 V. Capacitor, Mica, 240 mmf., 5%, 500 V. Capacitor, Paper, .01 mfd., 600 V. Capacitor, Paper, .047 mfd., 600 V. Capacitor, Paper, .01 mfd., 600 V.
If a station is transmitting a continuous-tone modulated signal, it may be used	C130	250175-3	Capacitor, Ceramic, 1500 mmf., 450 V.
		250159-56	Capacitor, Mica, 240 mmf., 5%, 500 V.
instead of the signal generator in the procedure just described. Keep in mind	C131		Capacitor, Paper, .01 mfd., 600 V.
the -5 V maximum reading on the VTVM.	C132	250203-7	Composition Peners Oliver Med 600 V
1	C133	250203-11	Capacitor, Paper, .047 mfd., 600 V.
This method may also be used to check the accuracy of a signal generator. Align	C134	250203-7	Capacitor, Paper, .01 mfd., 600 V.
the sound IF and ratio detector using an air signal, then substitute the signal	C135	250187-49	Capacitor, Mica, 47 mmf., 500 V.
generator as described in step 1. Adjust the frequency of the signal generator in	c1 36	250161-24	Capacitor, Mica, 4700 mmf., 55, 500 V.
he neighborhood of 4.5 mc. until the maximum reading on the VTVM is obtained.	C137	250203-13	Capacitor, Paper, .1 mfd., 600 V.
Tark that point on the signal generator dial as 4.5 mc.	C138	250203-11	Capacitor, Paper, .1 mfd., 600 V. Capacitor, Paper, .047 mfd., 600 V.
der v one horne on one prepar somerener crear en 4.5 mes	C139	250159 - 95	Capacitor, Mica, 33 mmf., 500 V.
	0T2A	とノシェノフ・ファ	
			MODELS CT270, CT271, CT272, CT273,
			CT274, CT275, CT276, CT277, CT278,
			CT279, CT280, CT281, CT282
			Camily Cambon Cambon Cambon

TIME	$\mathbf{m} \in \mathcal{O}_{\mathbf{m}}$	CHART

Measurements made with receiver operating on 117 volts 60 cycles AC, with a strong signal in put Measurements made with Voltohmist type VTVM, between indicated terminal and chassis ground.

TUBL NO.	TUBE	FUNCTION	PLATE PIN NO.	VOLTS		REEN VOLTS	CAT PIN NO.	HODE VOLTS		GRID VOLTS
V~101	SUAS	Sound IF	5	215	6	150	7	1.1	3.	0
V-102	6AL5	Sound Det.	7	-0.6			1	-0.4		
V-102	6AL5	Sound Det.	2 .	-0.3			5	0		
V-103A	ésn7ct	lst A.F.	2	2175			3	0	1	"- 0.7
V-103B	ésn7GT	Hor. AFC	5	195			6	- 8	4	-35
V-104	6V 6GT	Output'	3	368	4	392	8	158	5	140
V-105		Picture Tube	х	13KV	10	400	11	74	2	26
V-106	6AU6	lst Video IF	5	168	6	168	7	0.2	1	-3.8
V-107	6AU6	2nd Video IF	5	170	ć	170	7	0.18	1	-3.9
V-1C8	6AU6	3rd Video IF	5	168	6	168	7	1.55	1	О
V-109	6AL5	Video Det.	7	-3.1			1	0		
V-109	6AL5	AGC	2	4			5	1.55		
V-110	écbé	Video Amp.	5	135	6	132	7	2	1	1
VylllA	12AU7	S y nc, Amp.	1	130			3	0	2	-43
V- -111 B	12AU7	Sync.Sep.	6	3،3			8	81	7	0
V-112	604	Vert. Osc.	5	200			7	0	6	-81
V-1.13	6V 6GT	Vert.Output	3	378	4	378	3	11.5	5	- 24 5
V11.4A	6SN7GT	Sync, Clipper	5	47			6	0	4	-1.1
V1148	6SN7GT	Hor. Osc.	2	225			3	0	1	-94
V115	€BG €G	Hor. Output	Do not meas.	, , , , , , , , , , , , , , , , , , , 	દ	290	3	2.6	5	-26
V- 117	ówągt	Damper	Do not meas.				3	600		

PF tube voltages vary with AGC.

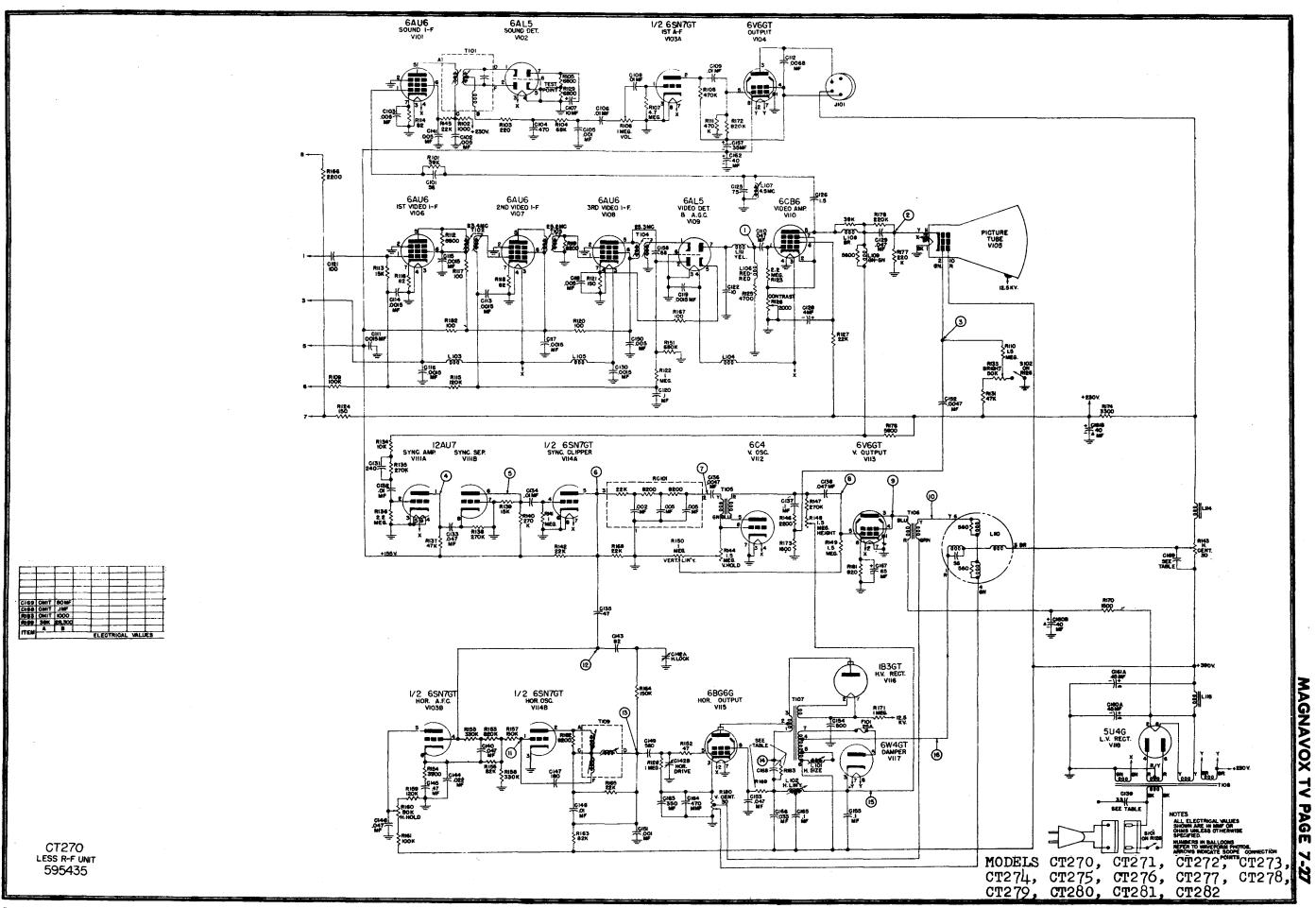
Osc. and hixer voltages vary with switch setting.

Video Ampl. plate and screen varies with different tubes.

Do not measure Horiz. Output plate or Damper plate with VTVM. Pulses cause grid rectification in meter so reading is meaningless and HV pulses may damage meter.

Ref. No.	Part No.	<u>Description</u>
T1 L1 L4 C1 C2 C3 C4 C5 C6 C7 C9 C10 C12 C13 C16 C17 C18 C19 C20 C21 R1 R2 R3 R4 R5 R6 R9 R10 R12 R13 R15 R16	Part No. 700349-1 360490-2 360372-1 360468-1 250176-18 250176-18 250176-18 250176-18 250164-7 250164-8 250164-6 250176-18 230104-78 230104-78 230104-78 230104-78 230104-68 230104-68 230104-68 230104-68	RF Tuner Unit Antenna Input Transformer Insulated RF Choke Converter Coil Capacitor, ceramic, 15 mmf., 500 V. Capacitor, ceramic, 470 mmf., 500 V. Capacitor, ceramic, 1.5 mmf., 500 V. Capacitor, ceramic, 1.5 mmf., 500 V. Capacitor, ceramic, 1.5 mmf., 500 V. Capacitor, ceramic, 2.2 mmf., 500 V. Capacitor, ceramic, 2.2 mmf., 500 V. Capacitor, ceramic, 470 mmf., 500 V. Capacitor, ceramic, 470 mmf., 500 V. Capacitor, ceramic, 470 mmf., 500 V. Trimmer Capacitor, ceramic, 15 mmf., 500 V. Resistor, ceramic, 470 mmf., 500 V. Resistor, carbon, 22K ohm, 1/2 W. Resistor, carbon, 6800 ohm, 1 W. Resistor, carbon, 6800 ohm, 1 W. Resistor, carbon, 20K ohm, 1/2 W. Resistor, carbon, 20K ohm, 1/2 W. Resistor, carbon, 6800 ohm, 1 V. Resistor, carbon, 6800 ohm, 1/2 W. Resistor, carbon, 6800 ohm, 1/2 W. Resistor, carbon, 47K ohm, 1/2 W. Resistor, carbon, 3300 ohm, 1/2 W. Resistor, carbon, 2000 ohm, 1/2 W.
L101 L102 L103 L104 L105 L106 L107 L108 L109 L110 L111	360357-1 360358-1 360332-11 360372-1 360372-1 360443-2 360443-13 360443-14 360462-6 360443-12 320041-3	Horiz. Size Coil Horiz. Lin. Coil Filament Choke Filament Choke Filament Choke Peaking Coil 4.5 Mc. Trap Peaking Coil Peaking Coil Deflection Yoke Peaking Coil Filter Choke Filter Choke





Ref. No	Part No.	Description		No. Part No. Description	modifica	ssis described herein an	ed or added t		eries. S eve ral The com-
C140 C141	250203-11 250175 - 1	Capacitor, Paper, .047 mfd., 600 V. Capacitor, Ceramic, 5000 mmf., 450 V.	R125 R126	230104-70 Resistor, carbon, 4700 ohm, 1/2 W. 220076-42 Potentiometer, Contrast		series at this writing in			!
C141	260106-3	Trimmer Condenser	R128	230104-98 Resistor, Carbon, 1 Meghohm, 1/2 W.	**	PICTURE			EIGHT OF PICTURE
C143	250187-52	Capacitor, Mica, 82 mmf., 500 V.	R129	230094-179 Resistor, Carbon, 6800 ohm, 5%, 1/2	W-CHASSIS	S TUBE , TYPE	AUDIO AMPLI	FIERS	TUBE CENTER
C144	250203-9	Capacitor, Paper, .022 mrd., 600 V.	R131 R133	230105-82 Resistor, Carbon, 47K ohm, 1W. 220076-15 Potentiometer, Brightness	CT270	14" Rectangular	Self Contai	t mad	9 "
C145	250205~17	Capacitor, Paper, .047 mfd. 600 V.	R134	230104-74 Resistor, Carbon, 10K ohm, 1/2 W.	"271	n n		inec 1. Radio used	9
C146	250203-11	Capacitor, Paper, .047 mfd., 600 V. Capacitor, Mica, 180 mmf., 5%, 500 V.	_	230104-91 Resistor, Carbon, 270 K ohm, 1/2 W.	"272	16" "	Self Contag		9 1/2
C147 C148	250157 - 73	Capacitor, Mica, 10,000 mmf., 300 V.	R136	230104-102 Resistor, Carbon, 2.2 Meghohm, 1/2 W	· "273	11 11		d. Radio used	9 1/4
C146	250159-130	Capacitor, Mica, 560 mmf., 500 V.	R137	230104-82 Resistor, Carbon, 47K ohm, 1/2 W.	"274		Self Conta		11 1/4
	250175-1	Capacitor, Ceramic, 5000 mmf., 450 V.	R138	230104-91 Resistor, Carbon, 270K ohm, 1/2 W.	"275	n n	None needed	d. Radio used	11 1/4
C151	250160-64	Capacitor, Mica, 1000 mmf., 500 V.	R139	230104-76 Resistor, Carbon, 15K ohm, 1/2 W.	"276	14" Rectangular	Self Conta:		9 1/2
C152	250203 - 5	Capacitor, Paper, .0047 mfd., 600 V.	R140	230104-91 Resistor, Carbon, 270K ohm, 1/2 W. 230104-98 Resistor, Carbon, 1 Megholm, 1/2 W.	"277 "278	16" "	Self Conta:		10 1/8
C153		Capacitor, Paper, .047 mfd., 600 V.	R141 .R142	230105-78 Resistor, Carbon, 1 Megnoum, 1/2 W.	"278	11 11		d. Radio used	10 1/8 11 1/4
C154	250189-1	H.V. Condenser	R143	220076-9 Potentioneter, Horiz. Centering	~ "279 "280		Self Conta: Self Conta:		10 1/8
C155	250203 - 13 250203-4	Capacitor, Paper, .1 mfd., 600 V. Capacitor, Paper, .0033 mfd., 600 V.		220076-37 Potentiometer, Vertical Hold	"281			d. Radio used	10 1/8
C156 C157	2700203-4	Capacitor, Electrolytic	R145	230104-78 Resistor, Carbon, 22K ohm, 1/2 W.	"282		Self Conta:		11 1/4
C158	250207-13	Capacitor, Ceramic, 68 mmf., 500 V.	R146	230104-69 Resistor, Carbon, 3900 ohms, 1/2 W.					·
C160	270021-39	Capacitor, Electrolytic	R147	230104-91 Resistor, Carbon; 270K ohm, 1/2 W.		the different methods o			
C161	270021-39	Capacitor, Electrolytic	R148	220076-28 Potentiometer, Vertical Height		require different gauss		The following cl	nart identifies
C162	270027-15	Capacitor, Electrolytic	R149	230104-100 Resistor, Carbon, 1.5 Meghohm, 1/2 W 220076-41 Potentiometer, Vertical Linearity	• the tra	up required for each type	e of tube.		!
C163	270027-18	Capacitor, Electrolytic	R150 R151	230104-96 Resistor, Carbon, 680K ohm, 1/2 W.		Ultrachità			1
C164	250159-90	Capacitor, Mica, 470 mmf., 500 V. Capacitor, Paper, .1 mfd., 600 V.	R152	230104-46 Resistor, Carbon, 47 Ohm, 1/2 W.		TUBE		то	N TRAP
C165	250203-10	Capacitor, Paper, .033 mfd., 600 V.	R153	230104-97 Resistor, Carbon, 820K ohm, 1/2 W.	RECTANG	GULAR BRAND RMA	CODE	PART NO.	
C166 C167	250205-10	Capacitor, Electrolytic	R154	230104-69 Resistor, Carbon, 3900 ohm, 1/2 W.	14BP4	Nat. Video	, 0000	360492 - 2	
CTO			R155	230104-92 Resistor, Carbon, 330K ohm, 1/2 W.	14BP4 14BP4	Nat. Video Rauland		360492 - 2	
R101	230104-81	Resistor, carbon, 39K ohm, 1/2 W.	R156	230104-85 Resistor, Carbon, 82K ohm, 1/2 W.	146P4 14CP4		188	360492 - 2	
R102	230104-62	Resistor, Carbon, 1K ohm, 1/2 W.	R157	230104-88 Resistor, Carbon, 150K ohm, 1/2 W.	14CP4	Nat. Union	247	360492 - 2	
R103	230104-54	Resistor, Carbon, 220 ohm, 1/2 W.	R158	230104-92 Resistor, Carbon, 330K ohm, 1/2 W. 230104-87 Resistor, Carbon, 120K ohm, 1/2 W.	14CP4	Dumont	157	360492-2	Yellow
R104	230104-84	Resistor, carbon, 68K ohm, 1/2 W.	R159 W. R160		16KP4	General Electric	188	360492-2	
R105	230094-17	9 Resistor, carbon, 6800 ohm, 5%, 1/2 W Potentiometer, Volume	R161	230104-86 Resistor, Carbon, 100K ohm, 1/2 W.	16KP4	Sheldon	636	360492-1	
R106	550010jr-10/	6 Resistor, carbon, 4.7 megohm, 1/2 W.		230104-73 Resistor, Carbon, 8200 ohm, 1/2 W.	16TP4	Arcturus ·	3 60	360492 -1	
R107 R108 '*	230105-94	Resistor, carbon, 470 K ohm, 1 W.	R163	230105-85 Resistor, Carbon, 82K ohm, 1 W.	16TP4 16TP4	Dumont Hytron	157 210	360492-2 360492 - 1	
R100	230104-86	Resistor, carbon, 100K ohm, 1/2 W.	R164		16TP4 16TP4	Nat. Video	210	360492 - 1	
R110	230104-86	Resistor, carbon, 100K ohm, 1/2 W.	R165		16TP4	Rauland		360492 - 2	
R111	53000/1-553	Resistor, carbon, 470K ohm, 5%, 1/2 W	1. R166	230105-66 Resistor, Carbon, 2200 ohm, 1 W.	16 T P4	Remington Rand	283	360492-1	Blue
R112	230104-72	Resistor, carbon, 6800 ohm, 1/2 W.	KT6.	230104-50 Resistor, Carbon, 100 onm, 1/2 w.	16TP4	Sheldon	636	360492-1	Blue
R111 R112 R113 R114	230104-76	Resistor, carbon, 15K ohm, 1/2 W.	R168 R169		16TP4	Sylvania	312	360492-1	
:R114	230104-49	Resistor, carbon, 82 ohm, 1/2 W. Resistor, carbon, 120K ohm, 1/2 W.	R170		16RP4	American TV	551	360492 -1	
R115	230104-01	Resistor, carbon, 82 ohm, 1/2 W.	R171	230105-98 Resistor, Carbon, 1 Meghohm, 1 W.	16RP4	Eureka	~~^	360492-1 360492-2	
R116 R117	230104-50	Resistor, carbon, 100 ohm, 1/2 W.	R172	230091-229 Resistor, Carbon, 820K ohm, 5%, 1/2	W.16RP4 16RP4	Hytron Raytheon	210 280	360492-2	
R117	230105-49	Resistor, carbon, 82 ohm, 1/2 W.	R174	240021-18 Resistor, W.W. 3300 ohm, 10%, 10 W.	16RP4 16RP4	Thomas	200	360492-1	
R119	230104-73	Resistor, carbon, 8200 ohm, 1/2 W.	R176	230106-71 Resistor, Carbon, 5600 ohm, 2 W.	16RP4	Federal T&R	179	360492 - 2	
R120	230104-50	Resistor, carbon, 100 ohm, 1/2 W.	R177	230104-90 Resistor, Carbon, 220K ohm, 1/2 W.	17AP4	Dumont	157	360492-2	Yellow
R121	230104-52	Resistor, carbon, 150 ohm, 1/2 W.	R178		17AP4	Thomas Electr.		360492-1	Blue
R122	230104-98	Resistor, carbon, 1 megohm, 1/2 W.	R180 R181		17BP4	General Electric	183	360492-2	
R123 R124	230104-107	2 Resistor, carbon, 2.2 megohm, 1/2 W. Resistor, carbon, 150 ohm, 1/2 W.	R182		17BP4	Nat. Union	247	360492-2	Yellow
R124	230104-72	Resistor, carbon, 270,,		a Jewe ()		Round	Brand	Ion	Trap
						16gP4	All	3601	450-3 Yellow